

Power Electronics

Poly Phase Uncontrolled Rectifiers

Dr. Firas Obeidat

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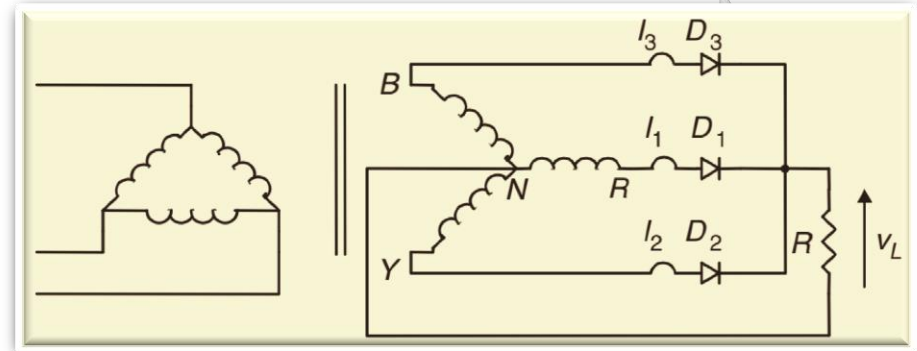
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Three Phase Uncontrolled Half Wave Rectifiers

A basic three-phase half-wave rectifier circuit with resistive load is shown in Figure. The rectifier is fed from an ideal 3-phase supply through delta-star 3-phase transformer.

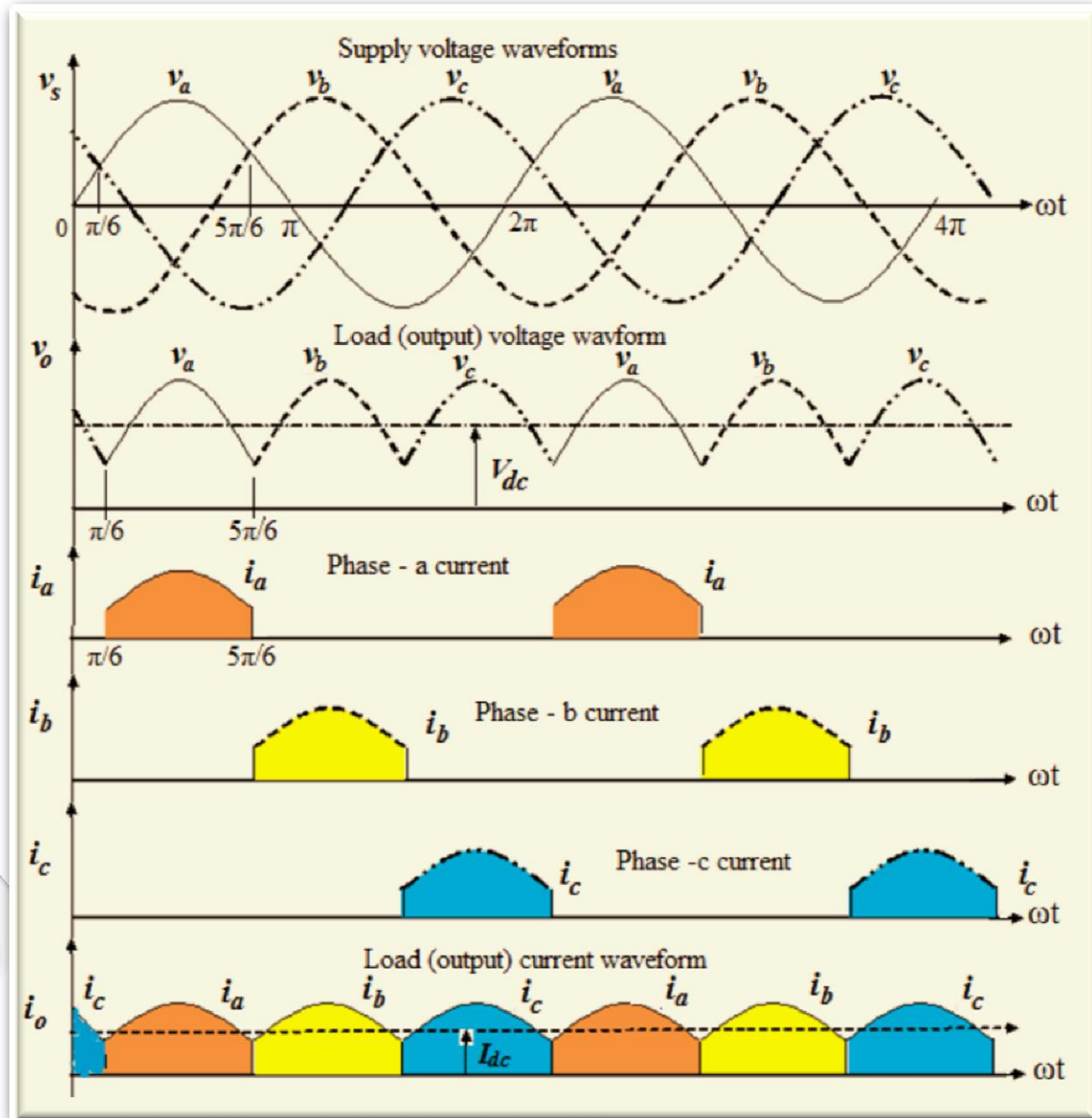


The principle of operation of this convertor can be explained as follows:

- The diode in a particular phase conducts during the period when the voltage on that phase is higher than that on the other two phases. For example: from $\pi/6$ to $5\pi/6$, D_1 has a more positive voltage at its anode, in this period D_2 and D_3 are off. The neutral wire provides a return path to the load current.
- The conduction sequence is: D_1, D_2, D_3 .

It is clear that, unlike the single-phase rectifier circuit, the conduction angle of each diode is $2\pi/3$, instead of π .

Three Phase Uncontrolled Half Wave Rectifiers



Three Phase Uncontrolled Half Wave Rectifiers

Variation of voltage across Diode D1

Voltage variation across diode D_1 can be obtained by applying KVL to the loop consisting of diode D_1 , Phase 'a' winding and load R.

$$\text{So, } -V_{D1} - V_o + V_a = 0 \quad \text{or} \quad V_{D1} = V_a - V_o$$

When Diode D1 conduct:

$$V_o = V_a$$

$$\text{Therefore, } V_{D1} = V_a - V_a = 0$$

When diode D_2 conduct:

$$V_o = V_b$$

$$\text{Therefore, } V_{D1} = V_a - V_b$$

$$\text{At } \omega t = 180^\circ, V_b = 0.866V_{mp}, V_a = 0$$

$$V_{D1} = -0.866V_{mp}$$

$$\text{At } \omega t = 210^\circ, V_b = V_{mp}, V_a = -0.5V_{mp}$$

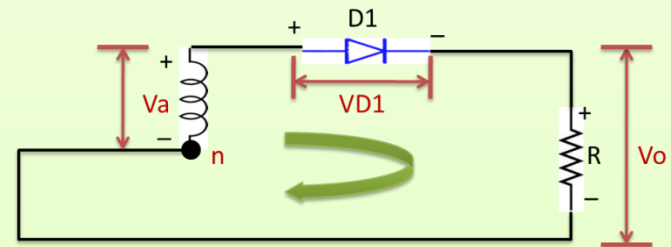
$$V_{D1} = -1.5V_{mp}$$

$$\text{At } \omega t = 240^\circ, V_b = 0.866V_{mp}, V_a = -0.866V_{mp}$$

$$V_{D1} = -\sqrt{3}V_{mp}$$

$$\text{At } \omega t = 270^\circ, V_b = 0.5V_{mp}, V_a = -V_{mp}$$

$$V_{D1} = -1.5V_{mp}$$



Three Phase Uncontrolled Half Wave Rectifiers

Variation of voltage across Diode D1

When Diode D3 conducts :

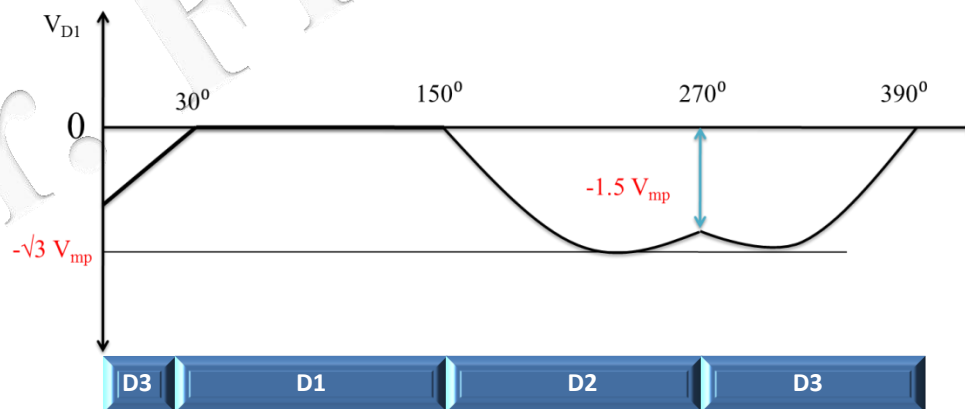
$$V_{D1} = V_a - V_c$$

$$\text{At } \omega t = 300^\circ, V_a = -0.866 V_{mp}, V_c = 0.866 V_{mp} \quad V_{D1} = -\sqrt{3} V_{mp}$$

$$\text{At } \omega t = 330^\circ, V_a = -0.5 V_{mp}, V_c = V_{mp} \quad V_{D1} = -1.5 V_{mp}$$

$$\text{At } \omega t = 360^\circ, V_a = 0, V_c = 0.866 V_{mp} \quad V_{D1} = -0.866 V_{mp}$$

$$\text{At } \omega t = 390^\circ, V_a = 0.5 V_{mp}, V_c = 0.5 V_{mp} \quad V_{D1} = 0$$



Three Phase Uncontrolled Half Wave Rectifiers

Let

$$V_{an} = V_m \sin \omega t$$

$$V_{bn} = V_m \sin(\omega t - 2\pi/3)$$

$$V_{cn} = V_m \sin(\omega t - 4\pi/3)$$

The **dc component** of the output voltage is the average value, and load current is the resistor voltage divided by resistance.

$$V_{dc} = \frac{3}{2\pi} \int_{\pi/6}^{5\pi/6} V_m \sin \omega t \, d\omega t = \frac{3\sqrt{3}V_m}{2\pi} = 0.827V_m$$

$$I_{dc} = \frac{3\sqrt{3}V_m}{2\pi R} = \frac{0.827V_m}{R}$$

The ***rms*** value of the output voltage and current are

$$V_{rms} = \sqrt{\frac{3}{2\pi} \int_{\pi/6}^{5\pi/6} (V_m \sin \omega t)^2 \, d\omega t} = 0.84V_m$$

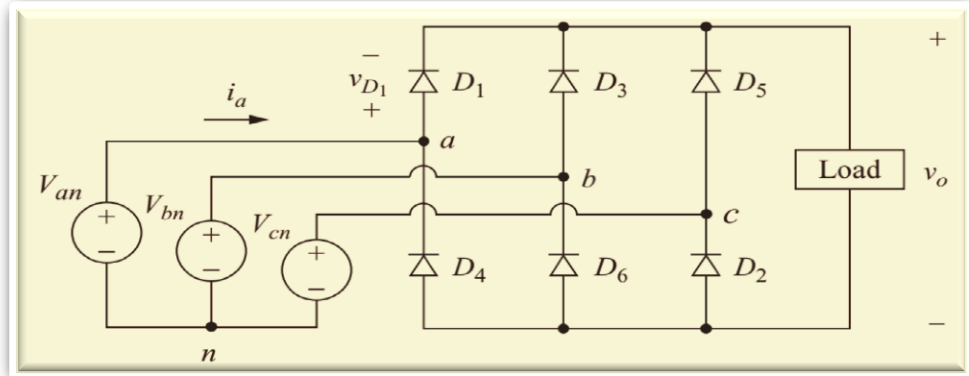
$$I_{rms} = \frac{0.84V_m}{R}$$

The ***rms*** current in each transformer secondary winding can also be found as

$$I_{rms} = \sqrt{\frac{1}{2\pi} \int_{\pi/6}^{5\pi/6} (I_m \sin \omega t)^2 \, d\omega t} = 0.485I_m$$

Three Phase Full-Wave Uncontrolled Bridge Rectifier

Three-phase rectifiers are commonly used in industry to produce a dc voltage and current for large loads. The three-phase voltage source is balanced and has phase sequence a-b-c.



Some basic observations about the circuit are as follows:

- Kirchhoff's voltage law around any path shows that only one diode in the top half of the bridge may conduct at one time (D_1 , D_3 , or D_5). The diode that is conducting will have its anode connected to the phase voltage that is highest at that instant.
- Kirchhoff's voltage law also shows that only one diode in the bottom half of the bridge may conduct at one time (D_2 , D_4 , or D_6). The diode that is conducting will have its cathode connected to the phase voltage that is lowest at that instant.
- D_1 and D_4 cannot conduct at the same time. Similarly, D_3 and D_6 cannot conduct simultaneously, nor can D_5 and D_2 .

Three Phase Full-Wave Uncontrolled Bridge Rectifier

Some basic observations about the circuit are as follows:

- The output voltage across the load is one of the line-to-line voltages of the source. For example, when D_1 and D_2 are ON, the output voltage is v_{ac} . Furthermore, the diodes that are ON are determined by which line-to-line voltage is the highest at that instant. For example, when v_{ac} is the highest line-to-line voltage, the output is v_{ac} .
- There are six combinations of line-to-line voltages (three phases taken two at a time). Considering one period of the source to be 360° , a transition of the highest line-to-line voltage must take place every $360^\circ/6=60^\circ$. Because of the six transitions that occur for each period of the source voltage, the circuit is called a six-pulse rectifier.
- The fundamental frequency of the output voltage is 6ω , where ω is the frequency of the three-phase source.

Three Phase Full-Wave Uncontrolled Bridge Rectifier

The figures shows the phase voltages and the resulting combinations of line-to-line voltages from a balanced three-phase source and the current in each of the bridge diodes for a resistive load.

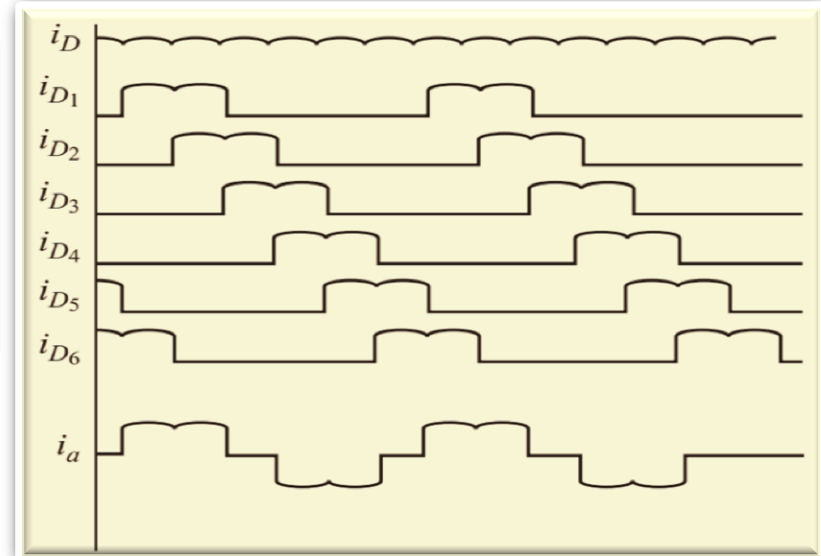
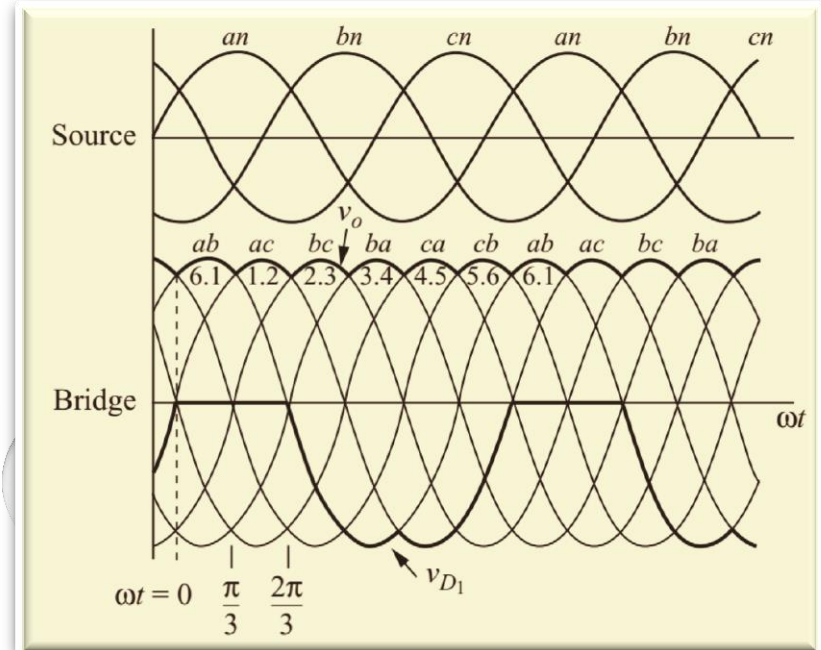
The diodes conduct in pairs (6,1), (1,2), (2,3), (3,4), (4,5), (5,6), (6,1), Diodes turn on in the sequence 1, 2, 3, 4, 5, 6, 1, . . .

The current in a conducting diode is the same as the load current. To determine the current in each phase of the source, Kirchhoff's current law is applied at nodes a, b, and c,

$$i_a = i_{D_1} - i_{D_4}$$

$$i_b = i_{D_3} - i_{D_6}$$

$$i_c = i_{D_5} - i_{D_2}$$



Three Phase Full-Wave Uncontrolled Bridge Rectifier

Let

$$V_{an} = V_m \sin \omega t$$

$$V_{bn} = V_m \sin(\omega t - 2\pi/3)$$

$$V_{cn} = V_m \sin(\omega t - 4\pi/3)$$

$$V_{ab} = V_{an} - V_{bn} = \sqrt{3}V_m \sin(\omega t + \pi/6)$$

$$V_{bc} = V_{bn} - V_{cn} = \sqrt{3}V_m \sin(\omega t - \pi/2)$$

$$V_{ca} = V_{cn} - V_{an} = \sqrt{3}V_m \sin(\omega t - 7\pi/6)$$

The dc component of the output voltage is the average value, and load current is the resistor voltage divided by resistance.

$$V_{dc} = \frac{3}{\pi} \int_{\pi/6}^{\pi/2} \sqrt{3}V_m \sin(\omega t + \pi/6) d\omega t = \frac{3}{\pi} \int_{\pi/6}^{\pi/2} \sqrt{3}V_m (\sin \omega t \cos \pi/6 + \cos \omega t \sin \pi/6) d\omega t$$

$$V_{dc} = \frac{3}{\pi} \int_{\pi/6}^{\pi/2} \sqrt{3}V_m \left(\frac{\sqrt{3}}{2} \sin \omega t + \frac{1}{2} \cos \omega t \right) d\omega t = \frac{3\sqrt{3}V_m}{\pi} = 1.654V_m$$

$$I_{dc} = 1.654 \frac{V_m}{R}$$

The average power is

$$P_{dc} = V_{dc} I_{dc}$$

Three Phase Full-Wave Uncontrolled Bridge Rectifier

The *rms* value of the output voltage is

$$V_{rms} = \sqrt{\frac{3}{\pi} \int_{\pi/6}^{\pi/2} (\sqrt{3}V_m \sin(\omega t - \pi/6))^2 d\omega t} = 1.655V_m$$

The *rms* current in each phase can also be found as

$$I_{(a,b,c)rms} = 0.78I_m$$

The *rms* current through a diode is:

$$I_{(D)rms} = 0.552I_m$$

Where

$$I_m = 1.73V_m/R$$

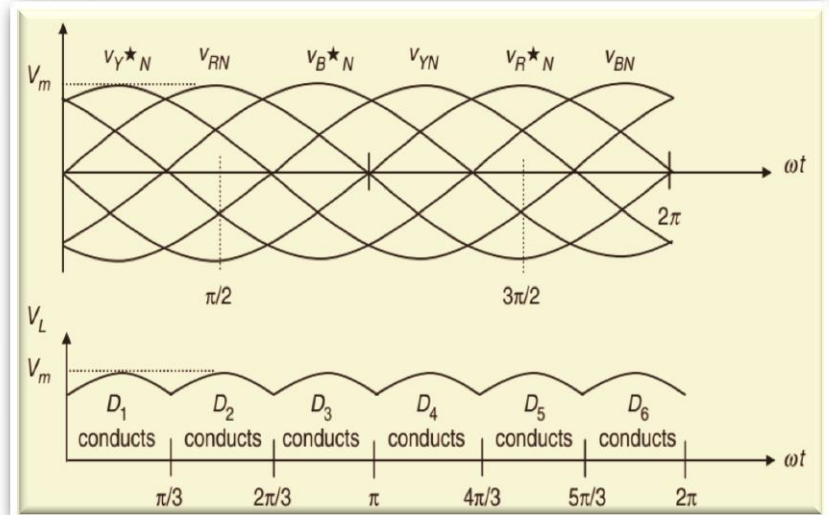
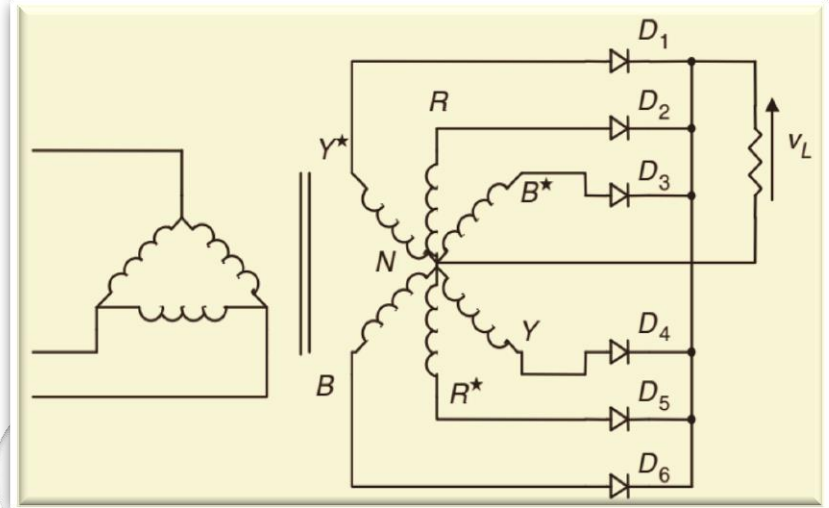
Six-phase Star Rectifier

The six-phase voltages on the secondary are obtained by means of a center-tapped arrangement on a star-connected three phase winding.

The diode in a particular phase conducts during the period when the voltage on that phase is higher than that on the other phases. The conduction angle of each diode is $\pi/3$.

Currents flow in only one rectifying element at a time, resulting in a low average current, but a high peak to an average current ratio in the diodes.

Six-phase star circuit is attractive in applications which require a low ripple factor and a common cathode or anode for the rectifiers.



Six-phase Star Rectifier

The average value of the output voltage can be found as

$$V_{dc} = \frac{6}{2\pi} \int_{\pi/3}^{2\pi/3} V_m \sin \theta d\theta = V_m \frac{6}{\pi} \frac{1}{2} = 0.955 V_m$$

The *rms* of the output voltage can be found as

$$V_{rms} = \sqrt{\frac{3}{\pi} \int_{\pi/3}^{2\pi/3} (V_m \sin \omega t)^2 d\omega t} = V_m \sqrt{\frac{3}{\pi} \left(\frac{\pi}{6} + \frac{\sqrt{3}}{4} \right)} = 0.956 V_m$$

The *rms* current in each transformer secondary winding can also be found as

$$I_{rms} = I_m \sqrt{\frac{1}{2\pi} \left(\frac{\pi}{6} + \frac{\sqrt{3}}{4} \right)} = 0.396 I_m$$

Where

$$I_m = \frac{V_m}{R}$$

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